



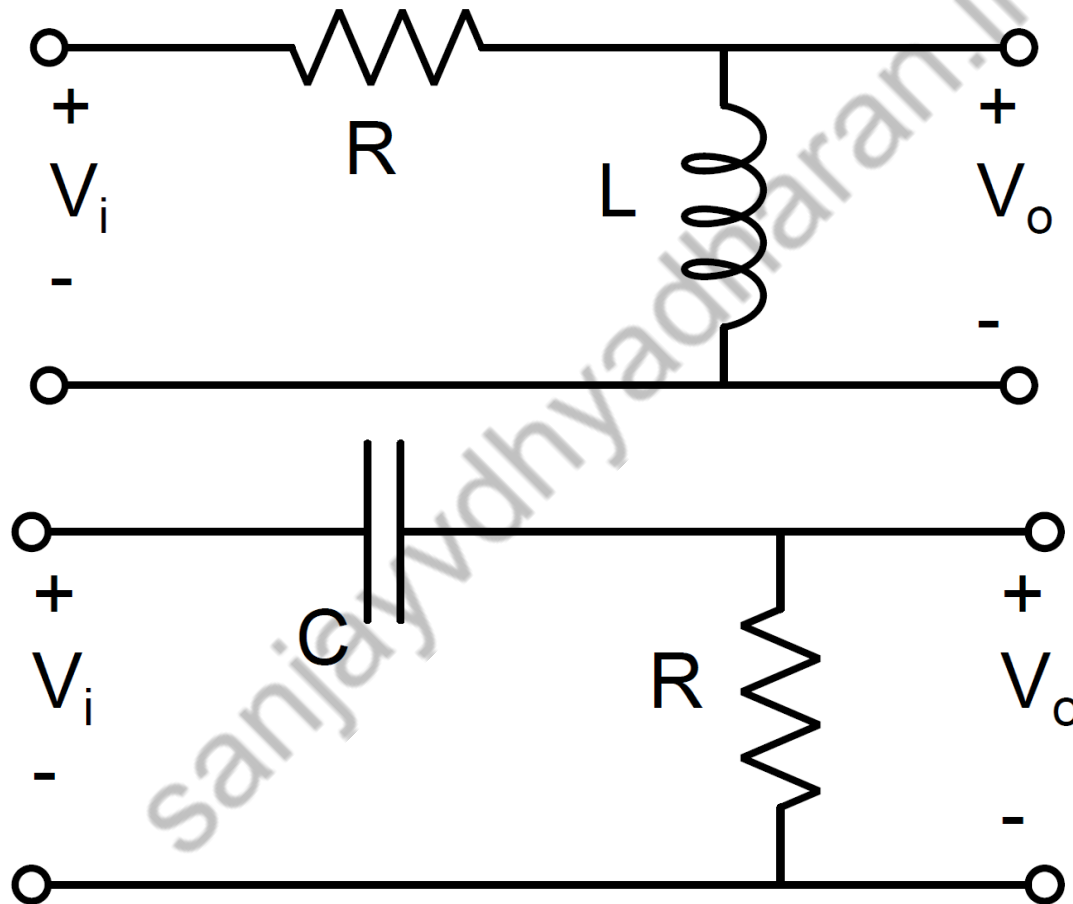
Electrical Science: 2021-22

Lecture 20

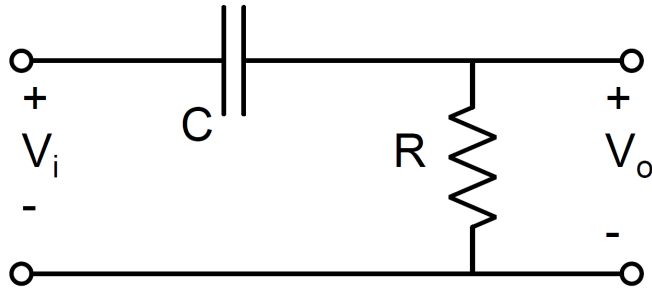
Passive Filters

By Dr. Sanjay Vidhyadharan

Passive High Pass Filters



Passive High Pass Filters



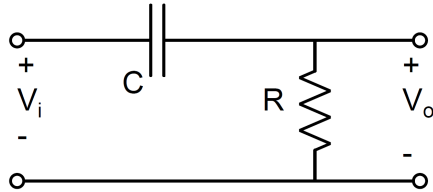
$$V_{out} = V_{in} \frac{R}{R - jX_C}$$

$$\text{Critical (or) Cutoff frequency } \omega_0 = \frac{1}{RC}$$

$$\text{Gain} = \frac{V_{out}}{V_{in}} = \frac{R}{R - jX_C} = \frac{1}{1 - j\frac{1}{RC\omega}} = \frac{1}{1 - j\frac{\omega_0}{\omega}}$$

$$\text{Gain}(j\omega) = \frac{1}{\sqrt{1 + \left(\frac{\omega_0}{\omega}\right)^2}} \angle \tan^{-1}\left(\frac{\omega_0}{\omega}\right)$$

Passive High Pass Filters



$$\omega_0 = \frac{1}{RC}$$

$$A_v = \text{Gain}(j\omega) = \frac{1}{\sqrt{1 + \left(\frac{\omega_0}{\omega}\right)^2}} \angle \tan^{-1}\left(\frac{\omega_0}{\omega}\right)$$

$$\text{Voltage Gain in dB} = 10 \log(A_v)$$

$$\text{Power Gain in dB} = 20 \log(A_v)$$

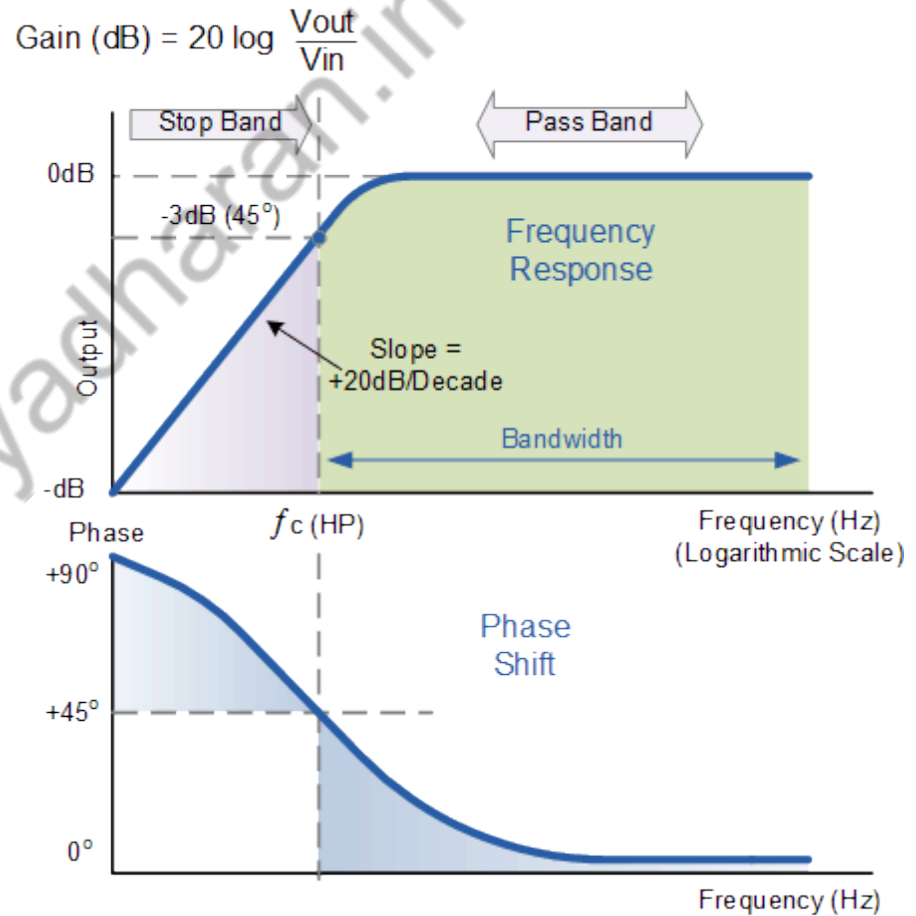
$$\text{Half Power Gain in dB} = 20 \log(0.5) = -3 \text{ dB}$$

$$\text{Half Power } A_v = \frac{1}{\sqrt{2}}$$

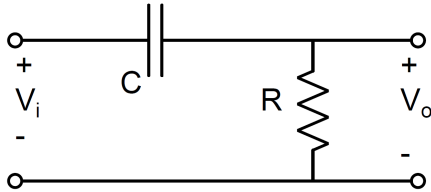
$$\text{For } \omega \gg \omega_0 \quad A_v = 1 \quad \text{Power Gain} = 0 \text{ dB}, \phi = 0$$

$$\text{For } \omega = \omega_0 \quad A_v = \frac{1}{\sqrt{2}} \quad \text{Power Gain} = -3 \text{ dB}, \phi = 45^\circ$$

$$\text{For } \omega < \omega_0 \quad \text{Slope } 20 \text{ dB/decade} \quad \text{for } \omega \ll \omega_0 \quad \phi = 90^\circ$$



Passive High Pass Filters



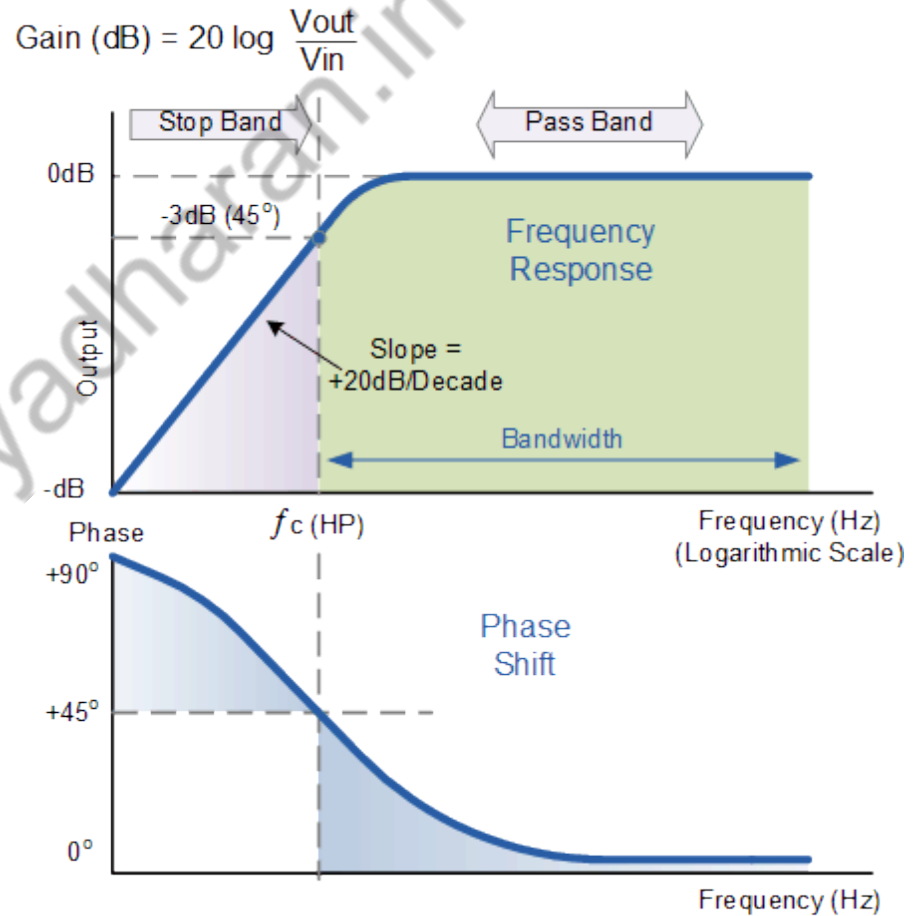
$$\omega_0 = \frac{1}{RC}$$

$$A_v = \text{Gain}(j\omega) = \frac{1}{\sqrt{1 + \left(\frac{\omega_0}{\omega}\right)^2}} \angle \tan^{-1}\left(\frac{\omega_0}{\omega}\right)$$

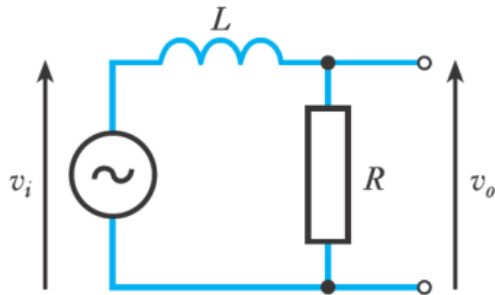
$$A_{v1} = \frac{1}{\sqrt{1 + \left(\frac{\omega_0}{\omega_1}\right)^2}}$$

$$A_{v2} = \frac{1}{\sqrt{1 + \left(\frac{\omega_0}{10\omega_1}\right)^2}} \approx 10 A_{v1}$$

$$\text{Gain in dB} = 20 \log 10 = 20 \text{ dB}$$



Passive High Pass Filters



$$\frac{v_o}{v_i} = \frac{Z_R}{Z_R + Z_L} = \frac{R}{R + j\omega L} = \frac{1}{1 + j\omega \frac{L}{R}} = \frac{1}{1 + j\frac{\omega}{\omega_c}} = \frac{1}{1 + j\frac{f}{f_c}}$$

At low frequencies, ω is small and the voltage gain is approximately 1.

At high frequencies, the magnitude of $\omega L/R$ becomes more significant and the gain of the network decreases.

$$|\text{voltage gain}| = \frac{1}{\sqrt{1 + \left(\omega \frac{L}{R}\right)^2}}$$

When the value of $\omega L/R$ is equal to 1, this gives

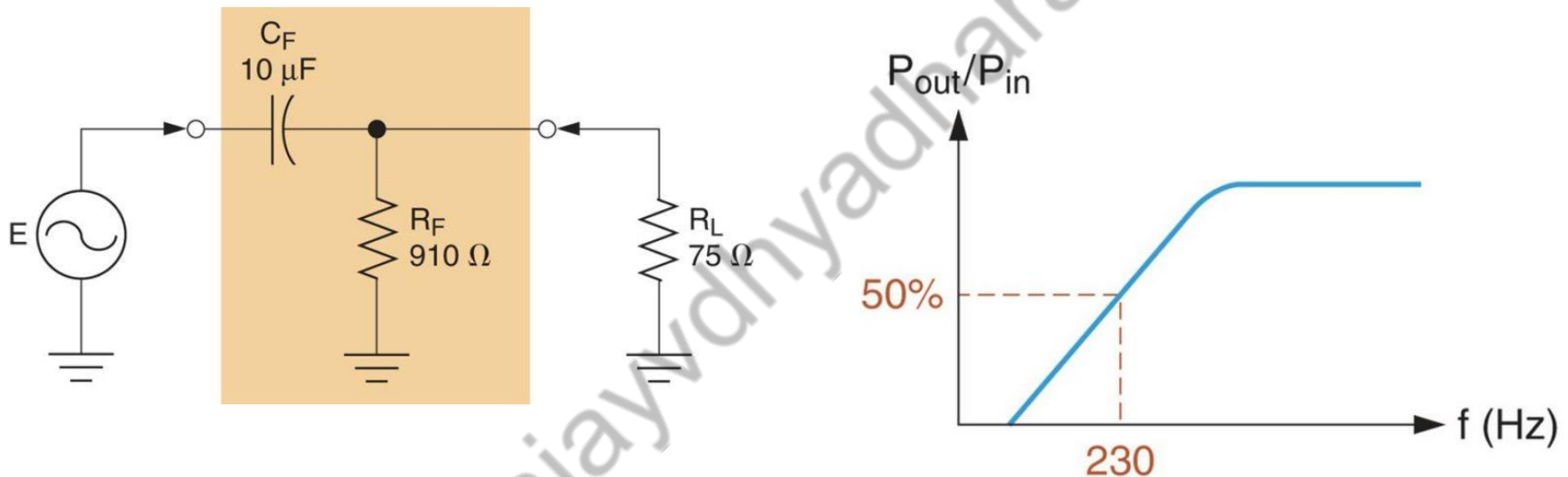
$$|\text{voltage gain}| = \frac{1}{\sqrt{1+1}} = \frac{1}{\sqrt{2}} = 0.707$$

$$A_v = \text{Gain}(j\omega) = \frac{1}{\sqrt{1 + \left(\frac{\omega}{\omega_0}\right)^2}} \angle \tan^{-1}\left(\frac{\omega_0}{\omega}\right)$$

$$\omega_0 = \frac{R}{L}$$

Passive High Pass Filters

Calculate the cutoff frequency for the RC High-Pass filter in Figure

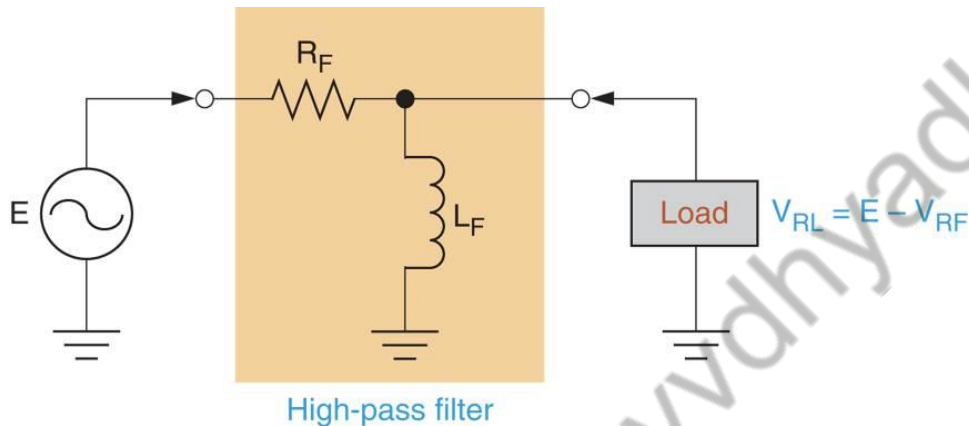


$$R_{EQ} = R_F || R_L = \frac{75 \times 910}{75 + 910} = 69.3 \Omega$$

$$f_c = \frac{1}{2\pi RC} = \frac{90.1 \Omega}{2\pi \times 69.3 \Omega \times 10 \mu\text{F}} = 230 \text{ Hz}$$

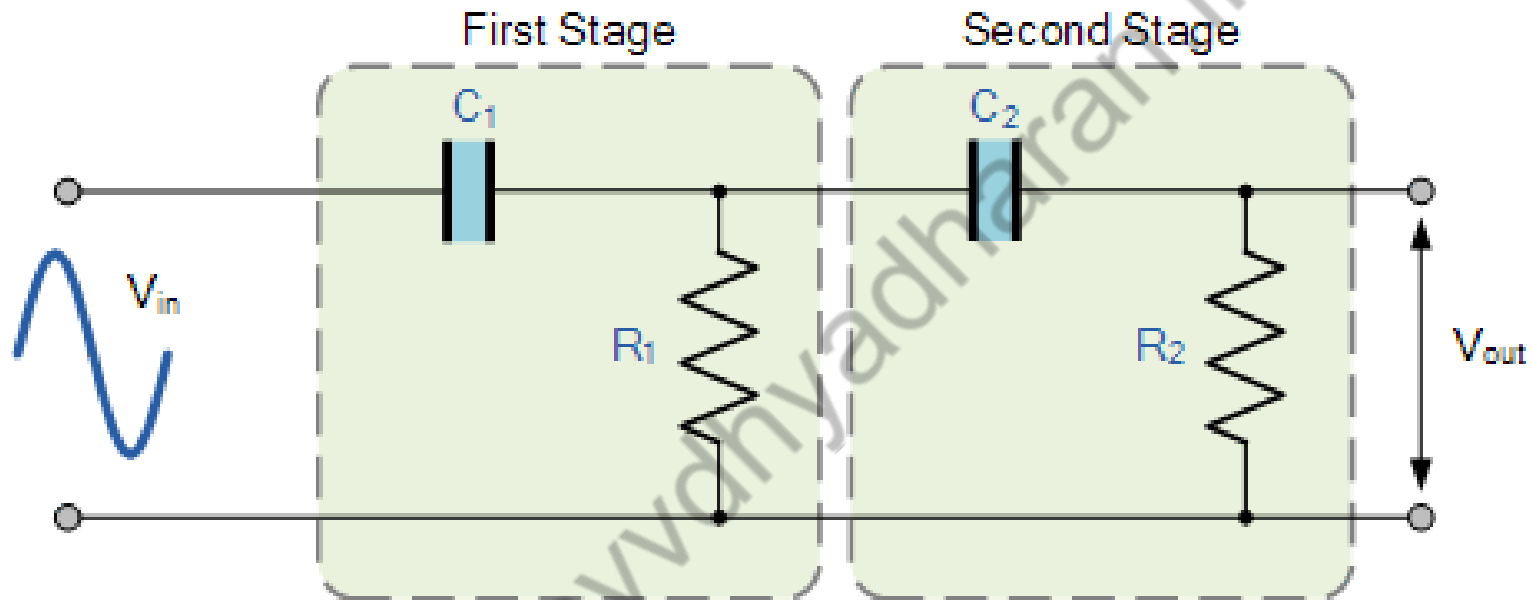
Passive High Pass Filters

Calculate the cutoff frequency for the RL High-Pass filter in Figure.
Resistor = 10KΩ, and Inductor = 470mH



$$f_c = \frac{R}{2\pi L} = \frac{10K\Omega}{2(3.14)(470mH)} = 3,388 \text{ Hz}$$

Second Order High Pass Filters

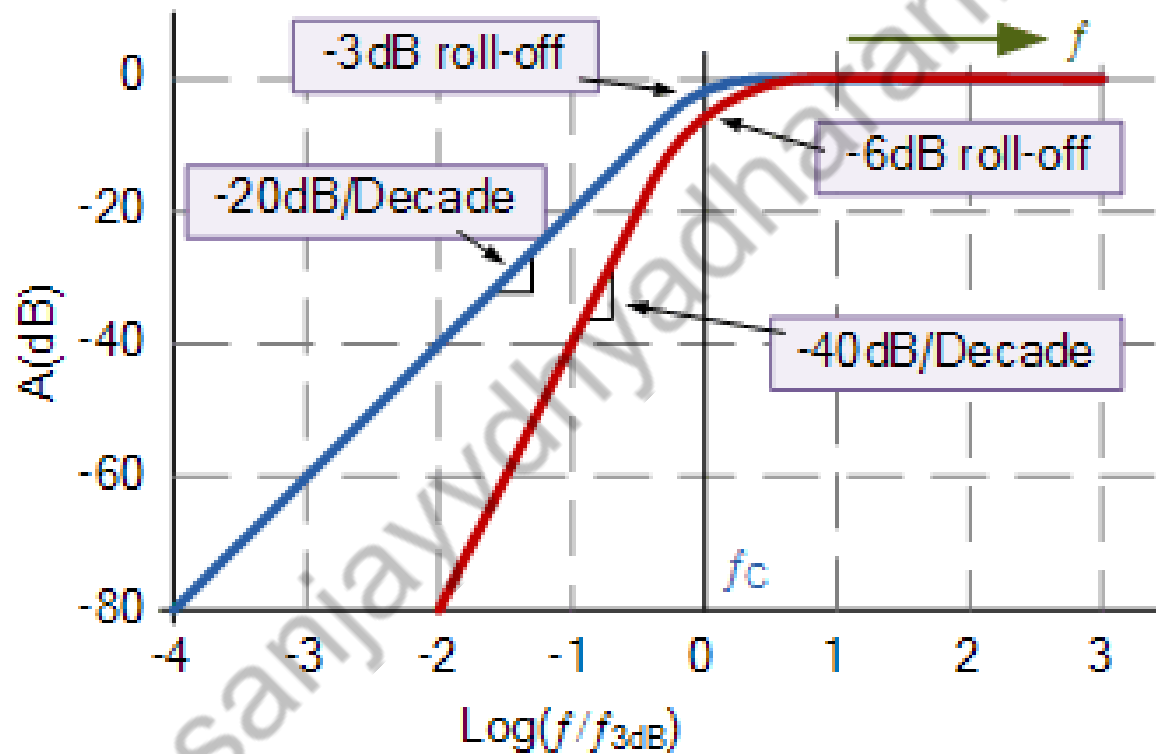


$$f_c = \frac{1}{2\pi \sqrt{R_1 C_1 R_2 C_2}} \text{ Hz}$$

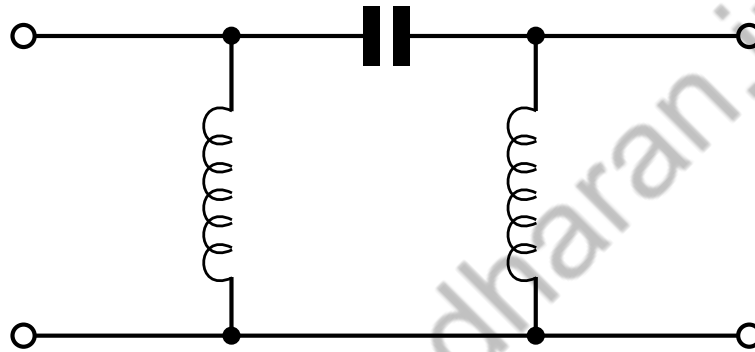
If Resistor and Capacitor values are the same:

$$f_c = \frac{1}{2\pi RC}$$

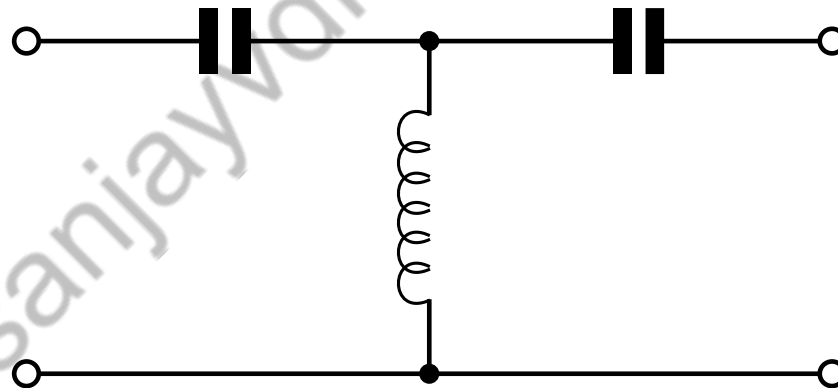
Second Order High Pass Filters



Other High Pass Filters

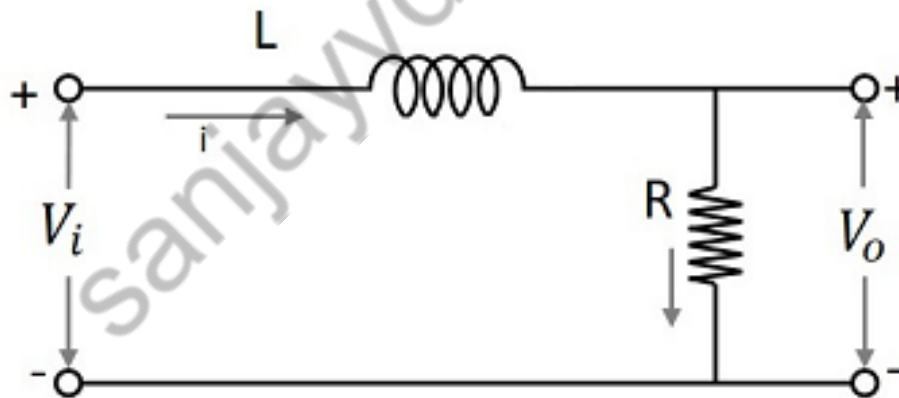
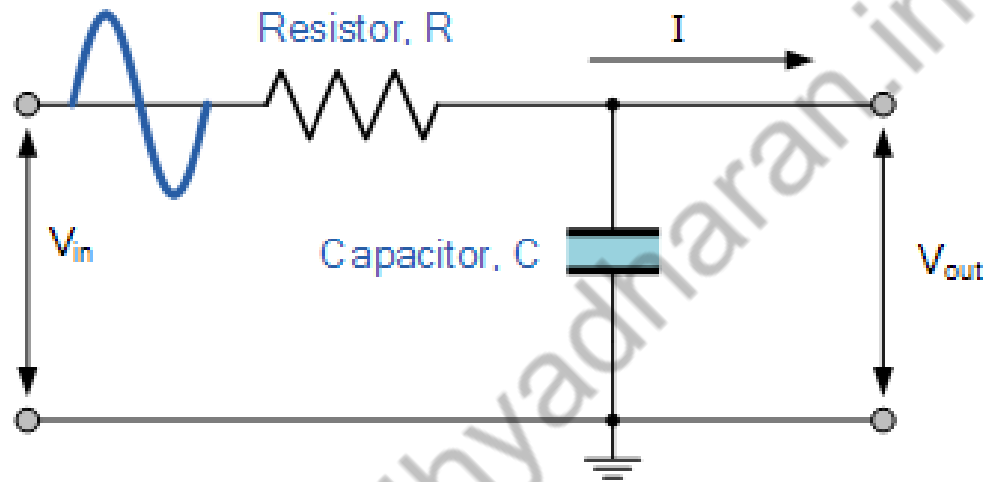


Topology for 3 pole π LC high pass RF filter



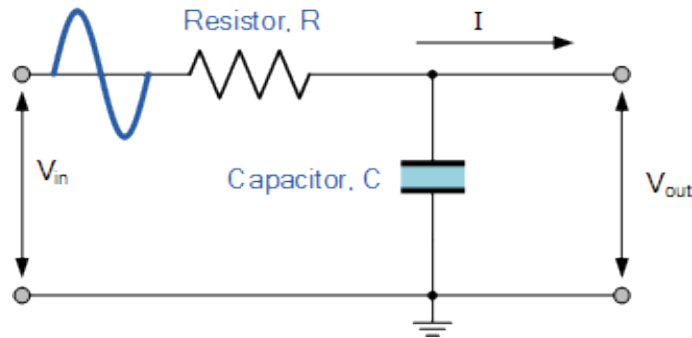
3 pole T LC high pass RF filter

Low Pass Filters



Low Pass RL Circuit

Low Pass Filters

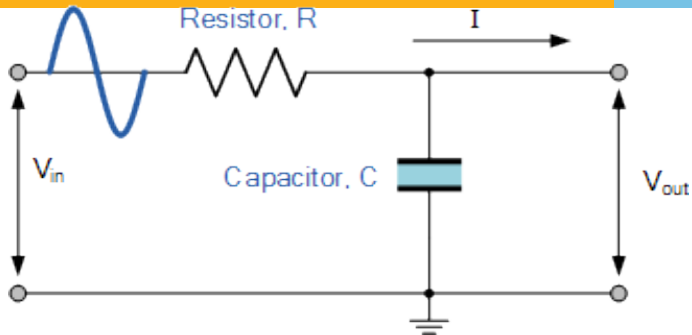


$$V_{out} = V_{in} \frac{-jX_c}{R - jX_c}$$

$$Gain = \frac{V_{out}}{V_{in}} = \frac{-jX_c}{R - jX_c} = \frac{1}{1 + jRC\omega} = \frac{1}{1 + j\frac{\omega}{\omega_0}} \quad \omega_0 = \frac{1}{RC}$$

$$Gain(j\omega) = \frac{1}{\sqrt{1 + \left(\frac{\omega}{\omega_0}\right)^2}} \angle -\tan^{-1}\left(\frac{\omega}{\omega_0}\right)$$

Low Pass Filters



$$\text{Gain}(j\omega) = \frac{1}{\sqrt{1 + \left(\frac{\omega}{\omega_0}\right)^2}} \angle -\tan^{-1}\left(\frac{\omega}{\omega_0}\right) \quad \omega_0 = \frac{1}{RC}$$

$$\text{Power Gain in dB} = 20 \log(A_v)$$

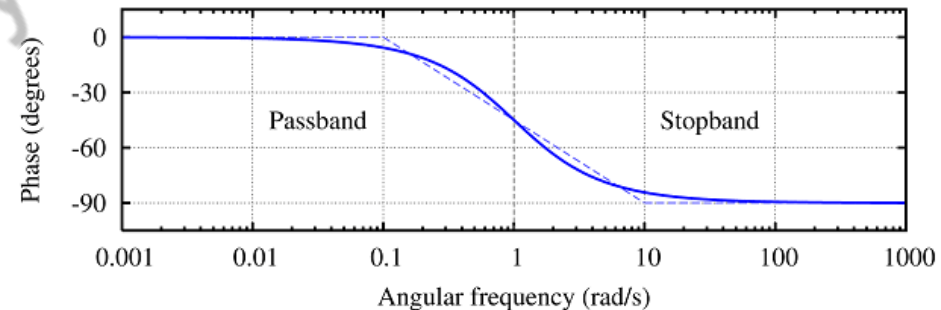
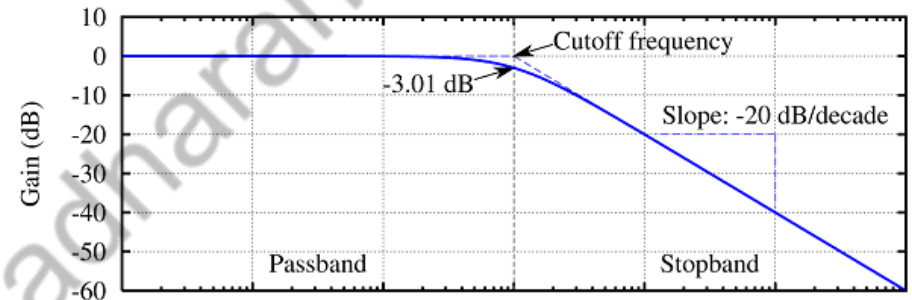
$$\text{Half Power Gain in dB} = 20 \log(0.5) = -3 \text{ dB}$$

$$\text{Half Power } A_v = \frac{1}{\sqrt{2}}$$

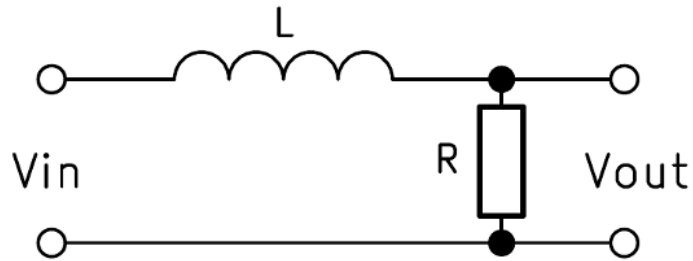
$$\text{For } \omega \ll \omega_0 \quad A_v = 1 \quad \text{Power Gain} = 0 \text{ dB}, \phi = 0$$

$$\text{For } \omega = \omega_0 \quad A_v = \frac{1}{\sqrt{2}} \quad \text{Power Gain} = -3 \text{ dB}, \phi = -45^\circ$$

$$\text{For } \omega > \omega_0 \quad \text{Slope } 20 \text{ dB/decade} \quad \text{for } \omega \gg \omega_0 \quad \phi = -90^\circ$$



Low Pass Filters



$$Gain(j\omega) = \frac{1}{\sqrt{1+(\frac{\omega}{\omega_0})^2}} \angle -\tan^{-1}(\frac{\omega}{\omega_0}) \quad \omega_0 = \frac{R}{L}$$

$$Power\ Gain\ in\ dB = 20 \log(A_v)$$

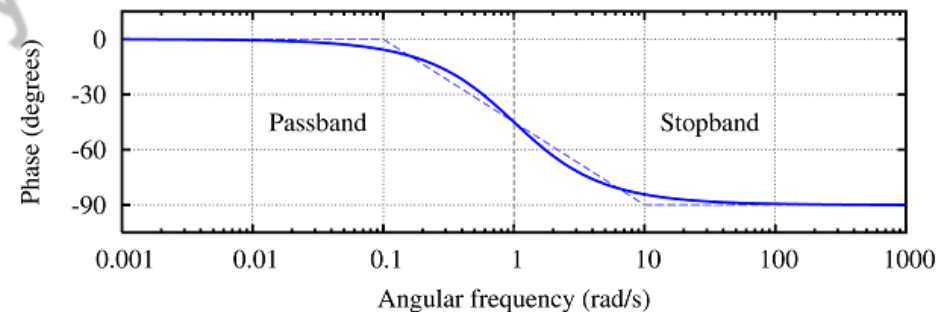
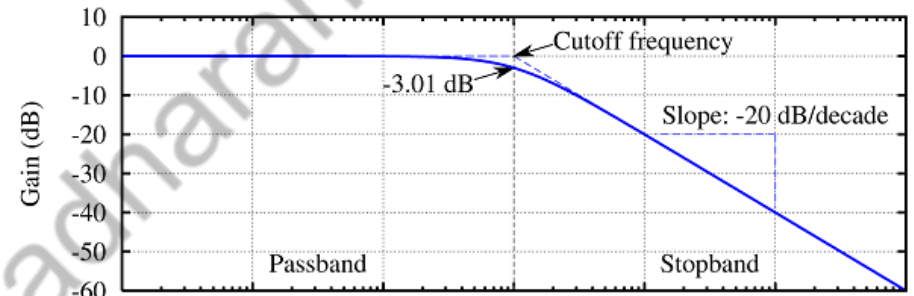
$$Half\ Power\ Gain\ in\ dB = 20 \log(0.5) = -3\ dB$$

$$Half\ Power\ A_v = \frac{1}{\sqrt{2}}$$

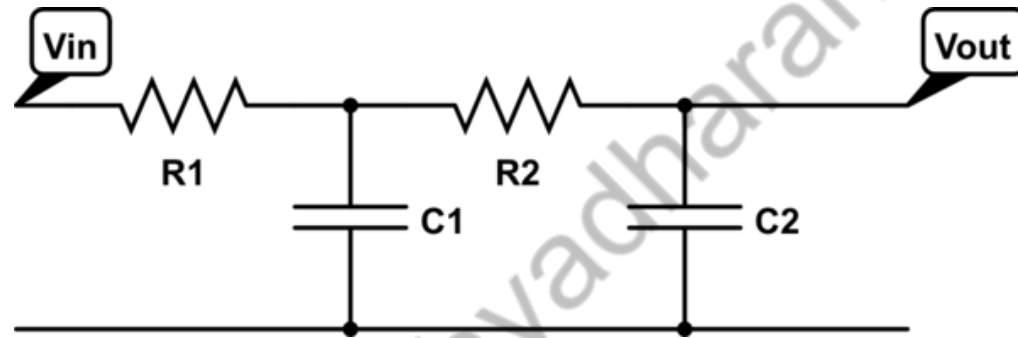
$$For\ \omega \ll \omega_0\ A_v = 1\ Power\ Gain = 0\ dB, \phi = 0$$

$$For\ \omega = \omega_0\ A_v = \frac{1}{\sqrt{2}}\ Power\ Gain = -3\ dB, \phi = -45^\circ$$

$$For\ \omega > \omega_0\ Slope\ 20\ dB/decade\ for\ \omega \gg \omega_0\ \phi = -90^\circ$$



Second Order Low Pass Filters

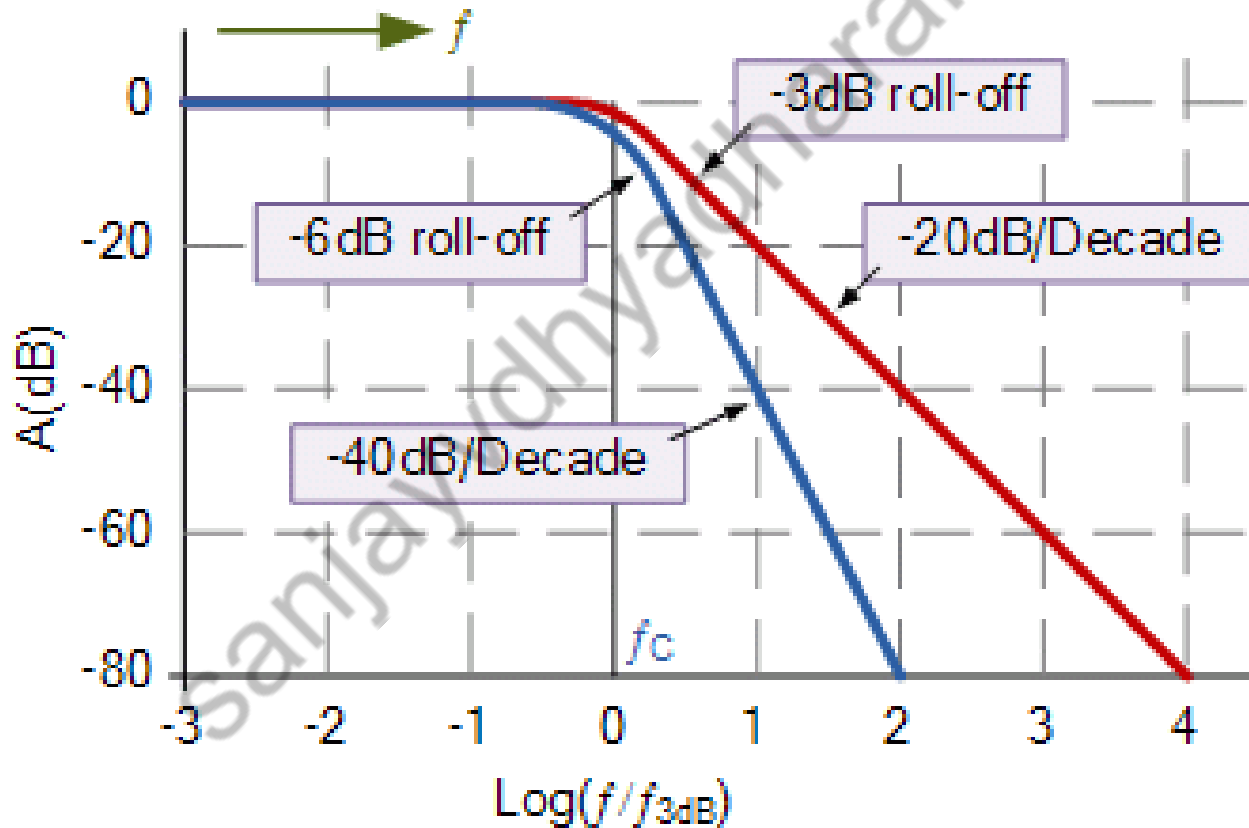


$$\omega_c = \frac{1}{\sqrt{R_1 C_1 R_2 C_2}}$$

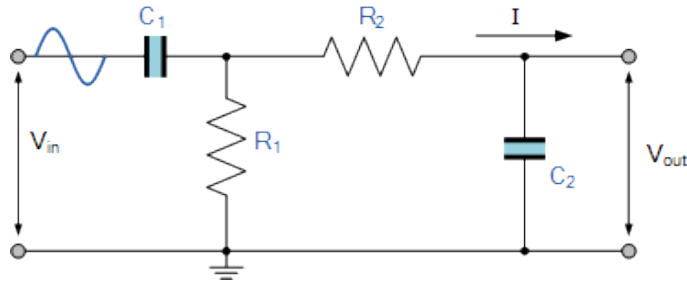
If Resistor and Capacitor values are the same:

$$f_c = \frac{1}{2\pi RC}$$

Second Order Low Pass Filters



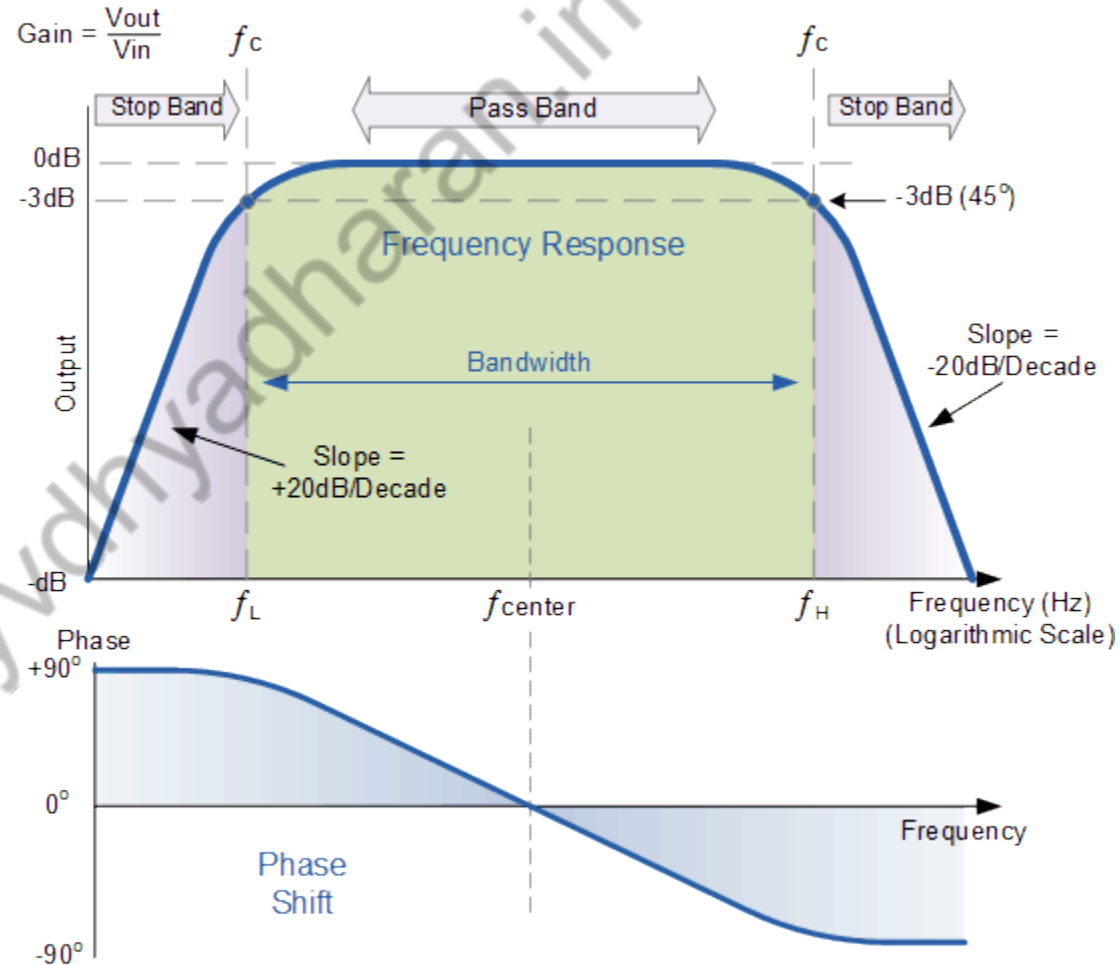
Band Pass Filters



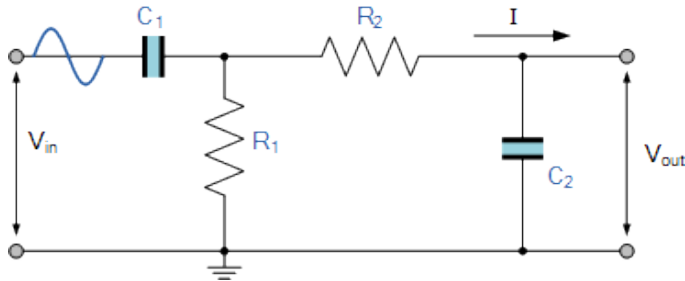
$$BW = f_H - f_L$$

Resonant Frequency (f_r)

$$f_r^2 = f_{(UPPER)} \times f_{(LOWER)}$$



Quality Factor of Band Pass Filters

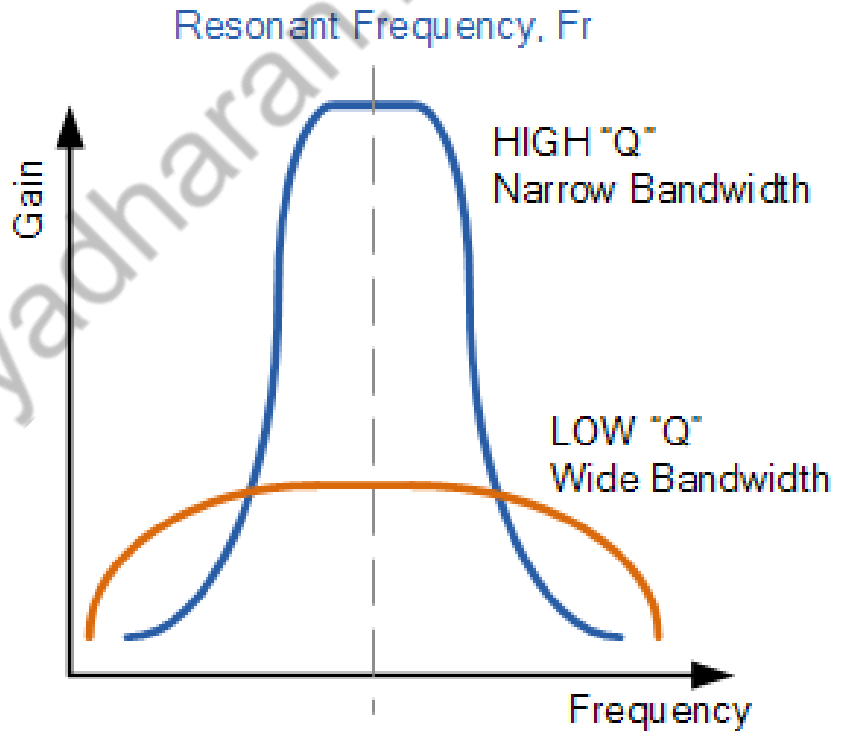


$$BW = f_H - f_L$$

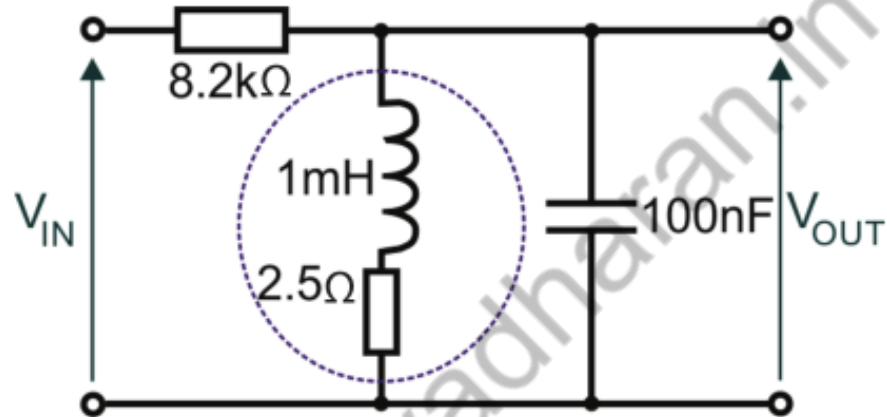
Resonant Frequency (f_r) or (f_0)

$$f_r^2 = f_{(UPPER)} \times f_{(LOWER)}$$

$$Q = \frac{f_o}{BW}$$



Band Pass Filters



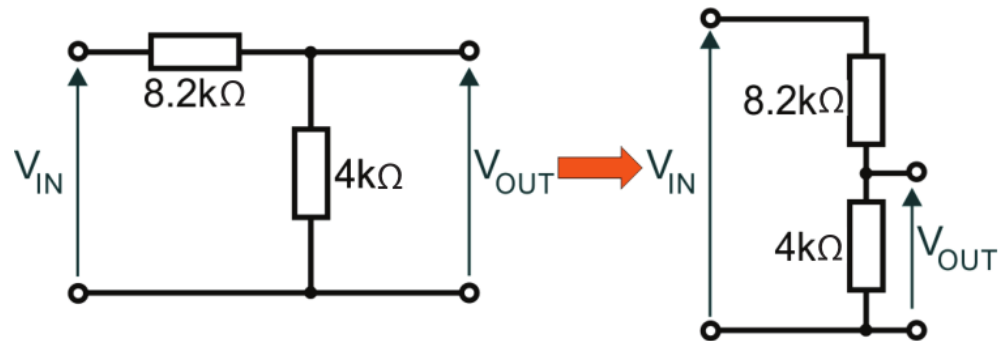
$$f_0 = \frac{1}{2\pi\sqrt{LC}}$$

$$= \frac{1}{2\pi\sqrt{1 \times 10^{-3} \times 100 \times 10^{-9}}}$$

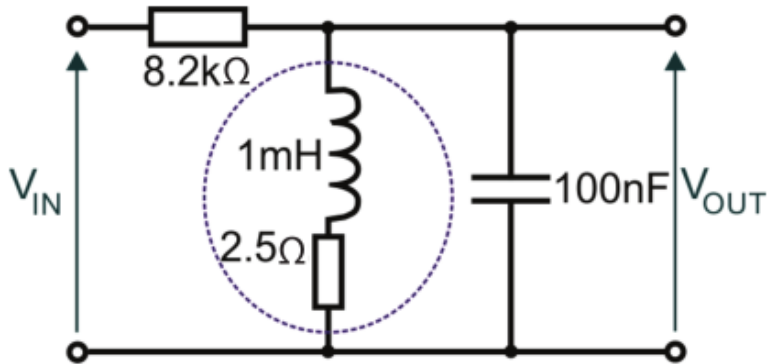
$$= 15915\text{Hz}$$

Dynamic resistance R_D

$$R_D = \frac{L}{r_L C} = \frac{1 \times 10^{-3}}{2.5 \times 100 \times 10^{-9}} = 4000\Omega$$



Band Pass Filters



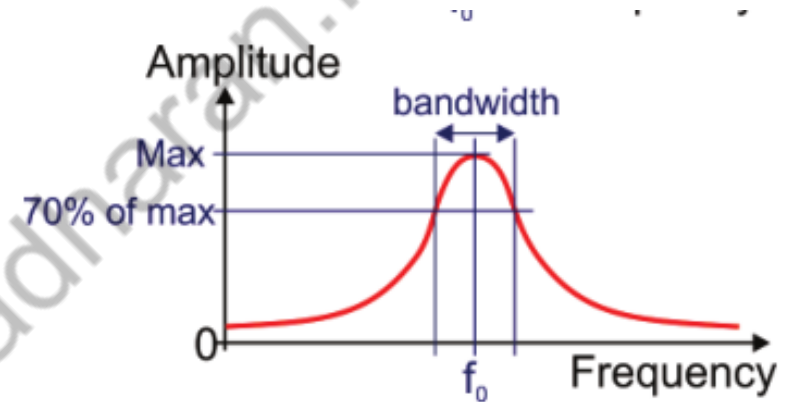
$$f_0 = \frac{1}{2\pi\sqrt{LC}}$$

$$= \frac{1}{2\pi\sqrt{1 \times 10^{-3} \times 100 \times 10^{-9}}}$$

$$= 15915\text{Hz}$$

$$R_D = \frac{L}{r_L C} = \frac{1 \times 10^{-3}}{2.5 \times 100 \times 10^{-9}} = 4000\Omega$$

$$BW = \frac{r_L}{L}$$



$$Q = \frac{2\pi f_0 L}{r_L} = \frac{2 \times \pi \times 15915 \times 1 \times 10^{-3}}{2.5} = 40$$

$$Q = \frac{f_0}{\text{bandwidth}} \rightarrow \text{Bandwidth} = \frac{f_0}{Q}$$

$$\text{Bandwidth} = \frac{f_0}{Q} = \frac{15915}{40} = 397.9\text{Hz}$$

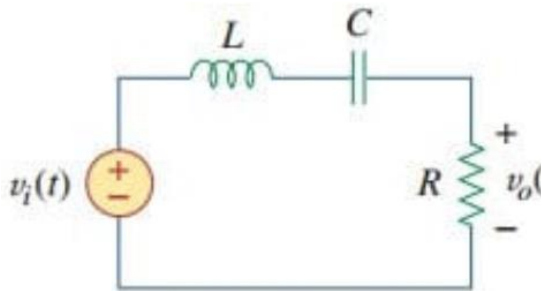
Band Pass Filters

The transfer function is

$$\mathbf{H}(\omega) = \frac{\mathbf{V}_o}{\mathbf{V}_i} = \frac{R}{R + j(\omega L - 1/\omega C)}$$

(5)

We observe that $\mathbf{H}(0) = 0$, $\mathbf{H}(\infty) = 0$. Figure.(7) shows the plot of $|\mathbf{H}(\omega)|$.



$$BW = R/L$$

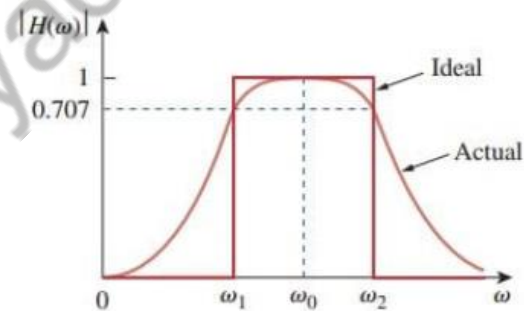
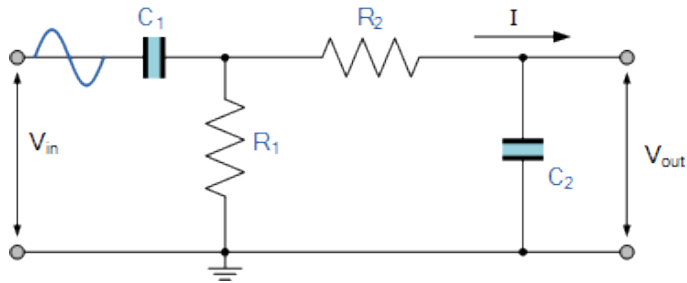


Figure 7. Ideal and actual frequency response of a bandpass filter.

The bandpass filter passes a band of frequencies ($\omega_1 < \omega < \omega_2$) centered on ω_0 , the center frequency, which is given by

$$\omega_0 = \frac{1}{\sqrt{LC}}$$

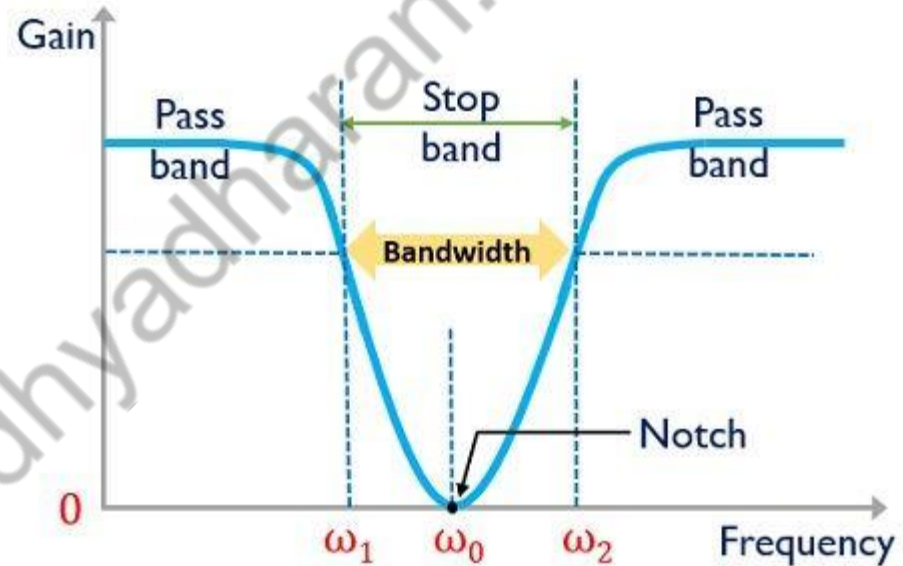
Band Reject Filters



$$BW = f_H - f_L$$

Resonant Frequency (f_r)

$$f_r^2 = f_{(UPPER)} \times f_{(LOWER)}$$



Practical Response

Electronics Coach

Band Reject Filters

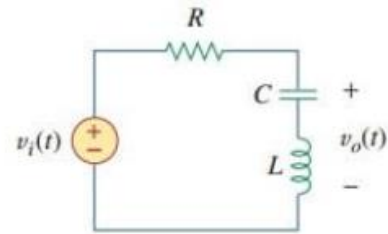


Figure 8. A bandstop filter.

$$\omega_0 = \frac{1}{\sqrt{LC}}$$

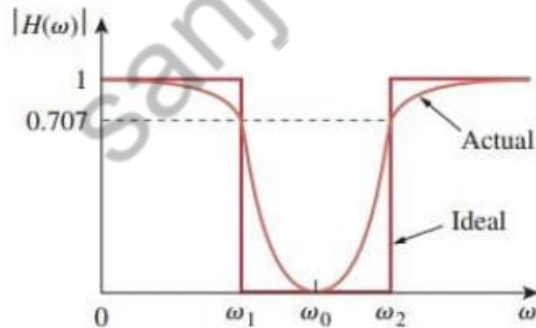
The transfer function is

$$\mathbf{H}(\omega) = \frac{\mathbf{V}_o}{\mathbf{V}_i} = \frac{j(\omega L - 1/\omega C)}{R + j(\omega L - 1/\omega C)}$$

$$BW = R/L$$

(7)

Notice that $\mathbf{H}(0) = 1$, $\mathbf{H}(\infty) = 1$. Figure 14.38 shows the plot of $|\mathbf{H}(\omega)|$.



Active Filters

Butterworth Filter

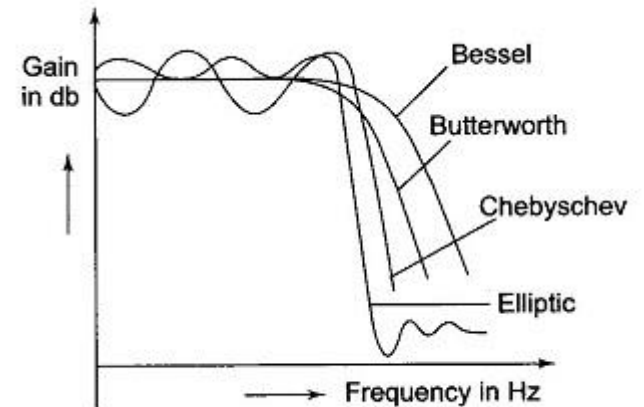
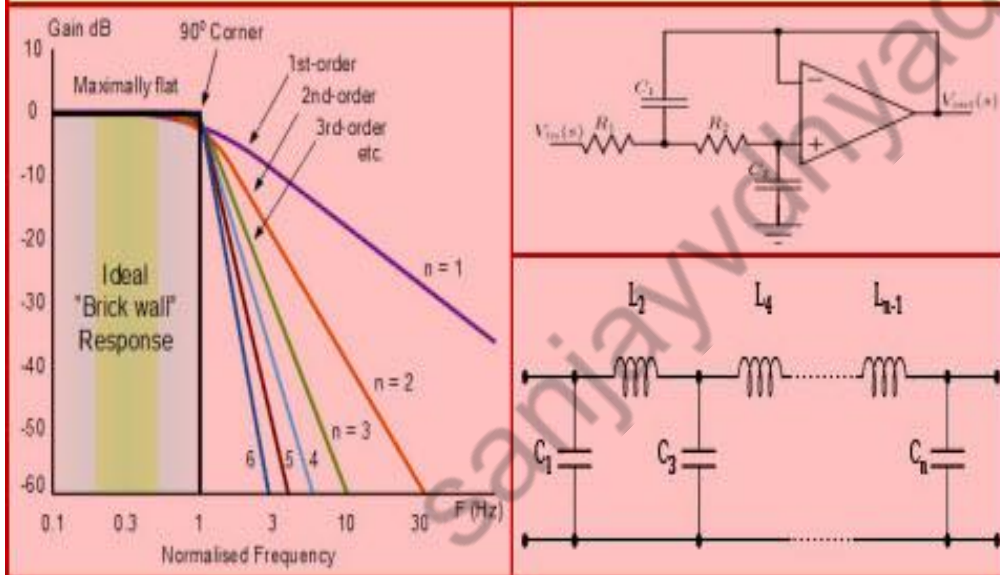


Fig. 15.27 Frequency Response of Various Filters

Thank you

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